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Sodankylä Geophysical Observatory, Finland

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Why KAIRA?

- Two years ago we wrote a proposal for a LOFAR remote station.
- The main goal was to try out the hardware as a EISCAT3D pathfinder.
- Since then, many more science goals have been attached to the project.

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What is LOFAR?



LEGO blocks for radio science





What is KAIRA?

LOFAR

- KAIRA is a dual array of wide band dipole VHF radio antennas
- A project of the Sodankylä Geophysical Observatory, principally funded by the University of Oulu in Finland.
- \blacktriangleright HBA: 30 \times 50 m radio telescope for the 120-240 MHz
- ▶ LBA: 30 m diameter radio telescope for the 30-80 MHz
- Uses proven LOFAR antenna and digital signal-processing hardware. Black sheep of LOFAR.
- Primary purpose to receive EISCAT VHF transmissions
- Multiple uses in geophysical remote sensing and radio astronomy
- Prototype for EISCAT3D



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LOFAR Antennas







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- Low Band Array 30 to 80 MHz.
- ▶ High Band Array 120 to 240 MHz.

LOFAR Antennas



LOFAR Antennas



Interferometry



KAIRA Site layout



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KAIRA Site layout



HBA Tile layout



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HBA Beam pattern





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What is a digital downconverter?

$$z_t = \sum_{n=1}^{N_{\rm int}} x_{n+tN_{\rm int}} \exp(i\omega_c(n+tN_{\rm int}))$$
(1)

What does FFT do?

$$z_{\omega} = \sum_{n=1}^{N} x_n \exp(i\omega n)$$
(2)

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Snow testing



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- Destructive testing
 - Snow impact, loading, lateral stresses, etc.

Snow testing continues...



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Ilkka's visit to Astron

Author: I.I. Virtanen	Date of issue: Kind of issue:	2012-Apr-13 Public	Scope: Project Documentation Doc.nr.: LOFAR-ASTRON-MAN-064	an.	
	Status:	Draft	File:		ΙΟΓΔΟ
	Revision nr.:	1.0		0	COLAR

Station Data Cookbook

Fast lag-profile inversion

- Lag-profile measurements are convolution measurements
- The transmission window in monostatic measurements prevents us from using frequency domain methods for deconvolution
- But bi-static observations have no transmission gaps and we can use FFT to dramatically speed up lag-profile inversion.

Lag-profile inversion

Consider the measurement equation for range and Doppler spread targets, but in this case using multiple different radar transmission envelopes indexed with c:

$$m_t^c = \sum_{r \in R} \epsilon_{t-r}^c \zeta_{r,t} + \xi_t.$$
(3)

If we now take conjugated self-products of these measurements with a lag $\tau.$ These can be organized as

$$m_t^c \overline{m_{t+\tau}^c} = \sum_{r \in R} \epsilon_{t-r}^c \overline{\epsilon_{t-r+\tau}^c} \sigma_r^\tau + \xi_t', \tag{4}$$

where $\sigma_r^{\tau} = E \zeta_{r,t} \overline{\zeta_{r,t+\tau}}$, and ξ'_t is a zero-mean noise term, which is dominated by the receiver noise in the case of low signal to noise ratio measurements. In the case of high signal to noise ratio measurements, this will also have significant zero mean contributions from the incoherent scatter cross-products $\zeta_{r,t} \overline{\zeta_{r',t+\tau}}$, where $r \neq r'$.

Lag-profile inversion

In more concise form, the lag-product equations can be stated as

$$m_t^{c,\tau} = \sum_{r \in R} \epsilon_{t-r}^{c,\tau} \sigma_r^\tau + \xi_t', \tag{5}$$

which is equivalent to the measurement equation for coherent (stationary) range-spread radar targets. For each lag τ , the measurement equations are different, as the ambiguity functions $\epsilon_{t-r}^{c,\tau}$ depend on the lag (and also transmission envelope). The equation is linear, i.e., the relationship between the unknown σ_r^{τ} and the measurements $m_t^{c,\tau}$ can be represented in matrix

$$\mathbf{m}^{\mathbf{c},\tau} = \mathbf{W}^{\mathbf{c},\tau}\boldsymbol{\sigma}^{\tau} + \boldsymbol{\xi}',\tag{6}$$

where the measurement vector $\mathbf{m}^{\mathbf{c},\tau}$ spans over all time indices that contribute to the unknown σ^{τ} .

Fast lag-profile inversion

$$\mathbf{x}_{\mathrm{ML}} = (\mathbf{A}^{\mathrm{H}} \boldsymbol{\Sigma}^{-1} \mathbf{A})^{-1} \mathbf{A}^{\mathrm{H}} \boldsymbol{\Sigma}^{-1} \mathbf{m}$$
(7)

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In frequency domain, the generalized linear least-squares for lag-profile inversion can be written as

$$\sigma_{r}^{\tau} = \mathcal{F}^{-1} \left\{ \frac{1}{\sum_{c=1}^{N_{\text{codes}}} |\epsilon_{c}^{\tau}(\omega)|^{2}} \sum_{c=1}^{N_{\text{codes}}} \overline{\epsilon_{c}^{\tau}}(\omega) m_{c}^{\tau}(\omega) \right\}$$
(8)

Bi-static configuration



First radar light 17.8.2012



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First coded pulse experiment 21.8.2012

2012-08-21 11:10:00.000 UT real part



Comparison with Tromso

Re(ACF) at 240 km



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Fast LPI result from all beams



Spectrum

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All-sky imaging



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All-sky imaging



Relative total electron content



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Beam shape using beacon satellites



Time

Wideband Imaging Riometer

» polarization, 39.062 MHz



- All-sky imaging can be piggy-backed on all LOFAR stations operating with the LBA antenna. This is done by using the cross-correlated signals from all baselines that can be stored in parallel with other LOFAR operations.
- It is also possible to perform a dedicated riometer experiment with beams tracking radio sources in the sky.
- Wide frequency range (30-80 MHz)

Wideband Imaging Riometer

What can we do with this?

- More tolerant to interference
- Accurate quite-day curve by tracking radio sources
- Inverting electron density profile from multi-frequency riometer absorption¹
- Inverting electron density and electron temperature from multi-frequency riometer absorption?

¹Lavergnat, J., and J. J. Berthelier (1973), An iterative mathematical technique for deriving electron-density profiles from multifrequency riometer data, Radio Sci., 8(7),

What next?

- Do more VHF measurements
- Develope imaging riometer software

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- Look at meteor head echos
- Look into IPS measurements



http://kaira.sgo.fi

